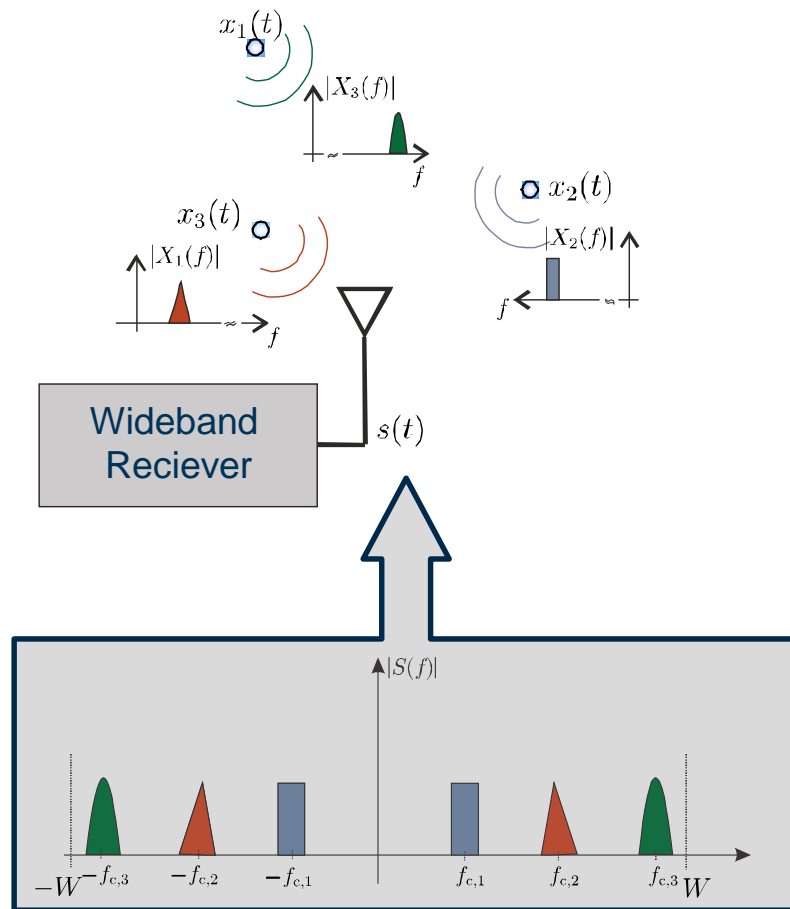


# ***Sub-Nyquist Wideband Acquisition and Spectrum Sensing of Multiband Signals***

2015 Wireless Innovation Forum  
Spectrum Sharing Workshop  
5<sup>th</sup>– 9<sup>th</sup> of October 2015, Erlangen, Germany

**Anastasia Lavrenko**, Florian Römer, Giovanni Del Galdo and Reiner Thomä

# Introduction



## Wideband Multiband Signal

- **few narrowband** communication **channels** distributed over **wide frequency band** of interest  $W$ .

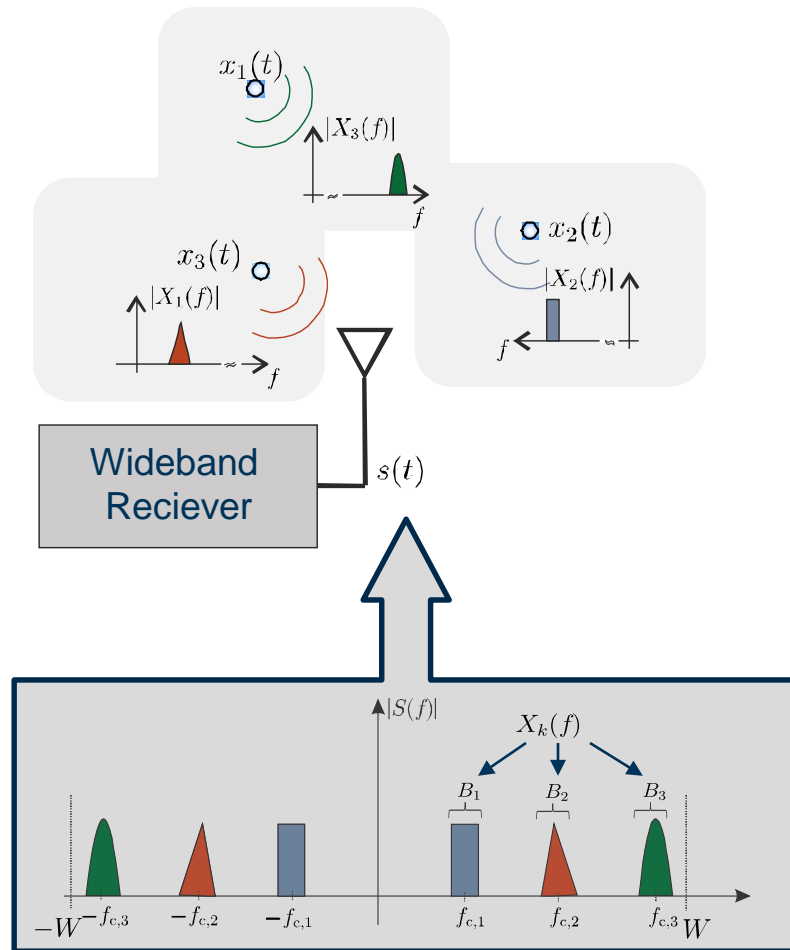
## Wideband Spectrum Sensing

- find out which **parts of spectrum** are **occupied** and which are not.

## Blind Wideband Spectrum Sensing

- **unknown** carrier frequencies

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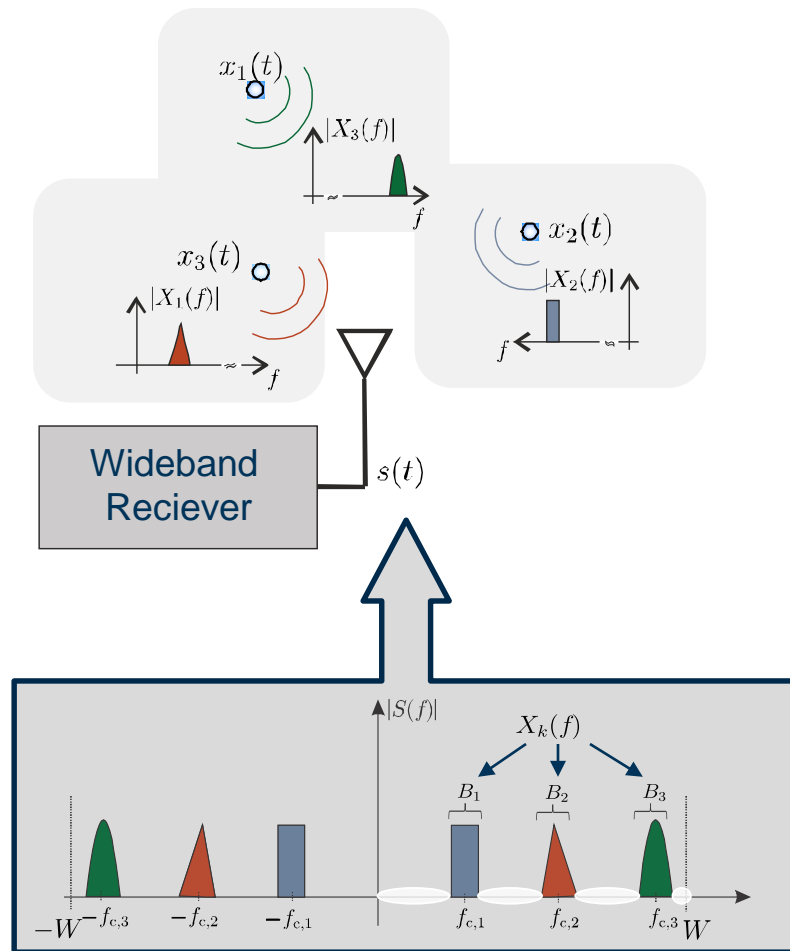
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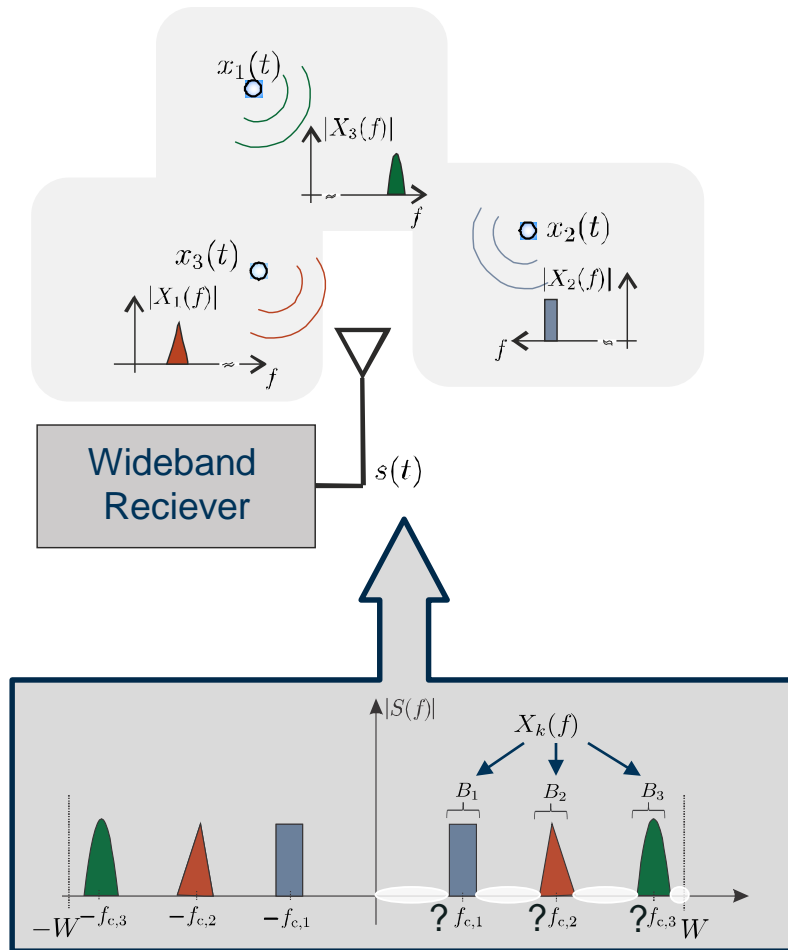
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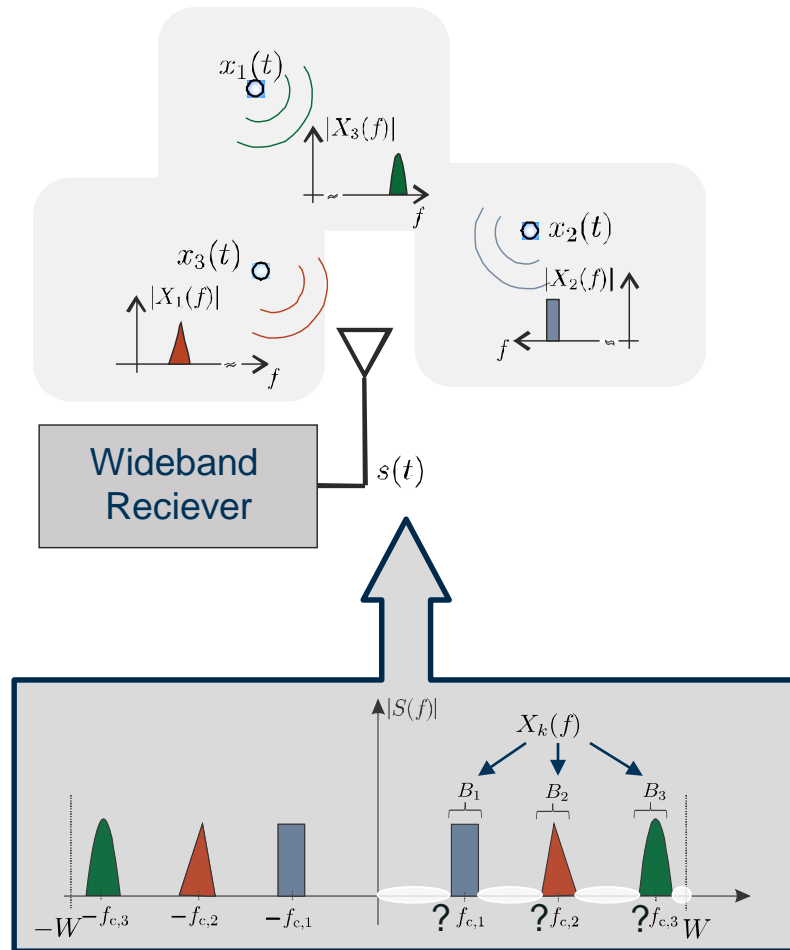
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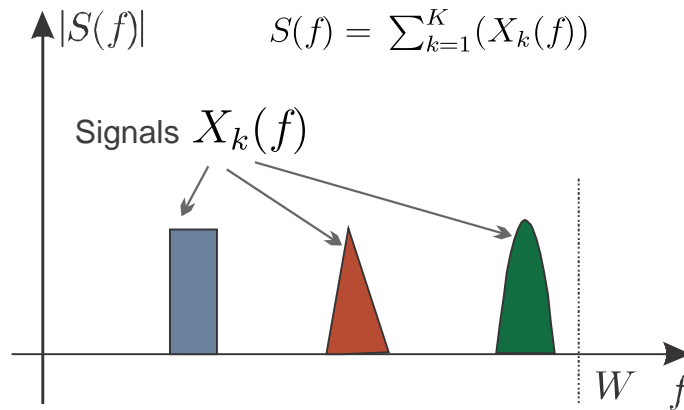
- **unknown** carrier frequencies

- Q1: can one **receive** such a wideband signal sampling it lower than required by Nyquist theorem?
- Q2: how can one **detect** spectral occupancy from the received time samples?
- Q3: how **accurate** this detection is?

# Outline

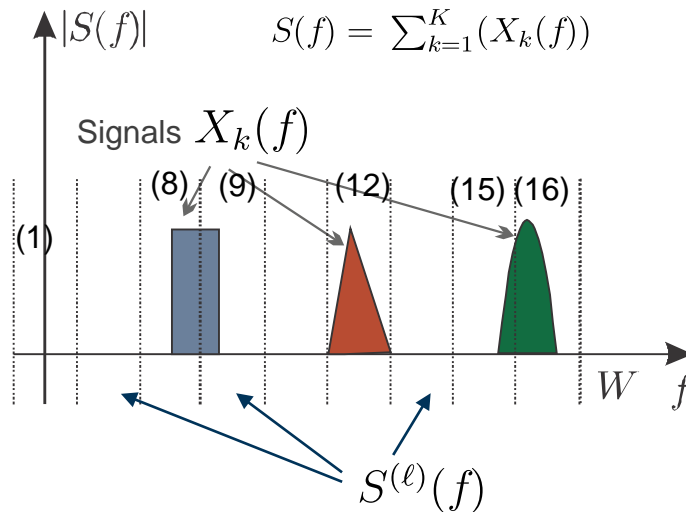
- Sub-Nyquist Signal Acquisition
  - main principle
  - examples of architectures
- Compressive Wideband Spectrum Sensing
  - Nyquist-sampled signal reconstruction
  - PSD reconstruction
  - direct coarse energy estimation
- Numerical examples

# Sub-Nyquist Signal Acquisition: Paradigm





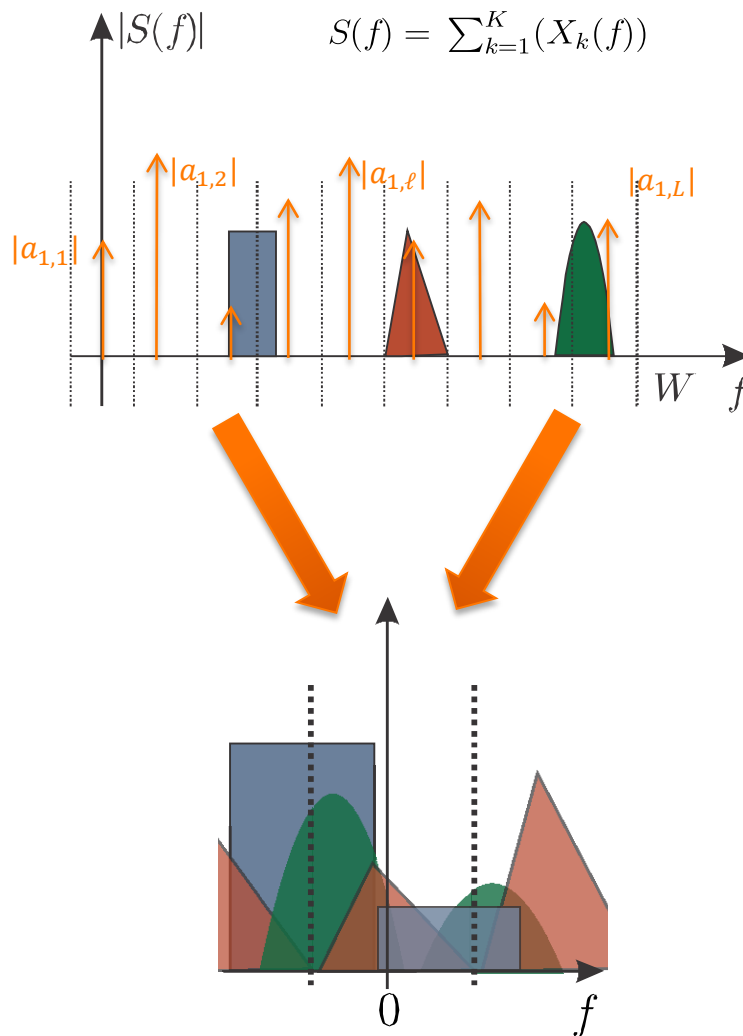
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- Virtually **split** the **Nyquist range** into  $L$  consecutive **non-overlapping blocks** (spectral cells) of width  $W/L$ .
- Treat **wideband signal**  $S(f)$  as a finite **union** of continuous **bandpass** (narrowband) **sub-spaces**

$$S(f) = [S^{(1)}(f), S^{(2)}(f), \dots, S^{(L)}(f)]$$

# Sub-Nyquist Signal Acquisition: Paradigm



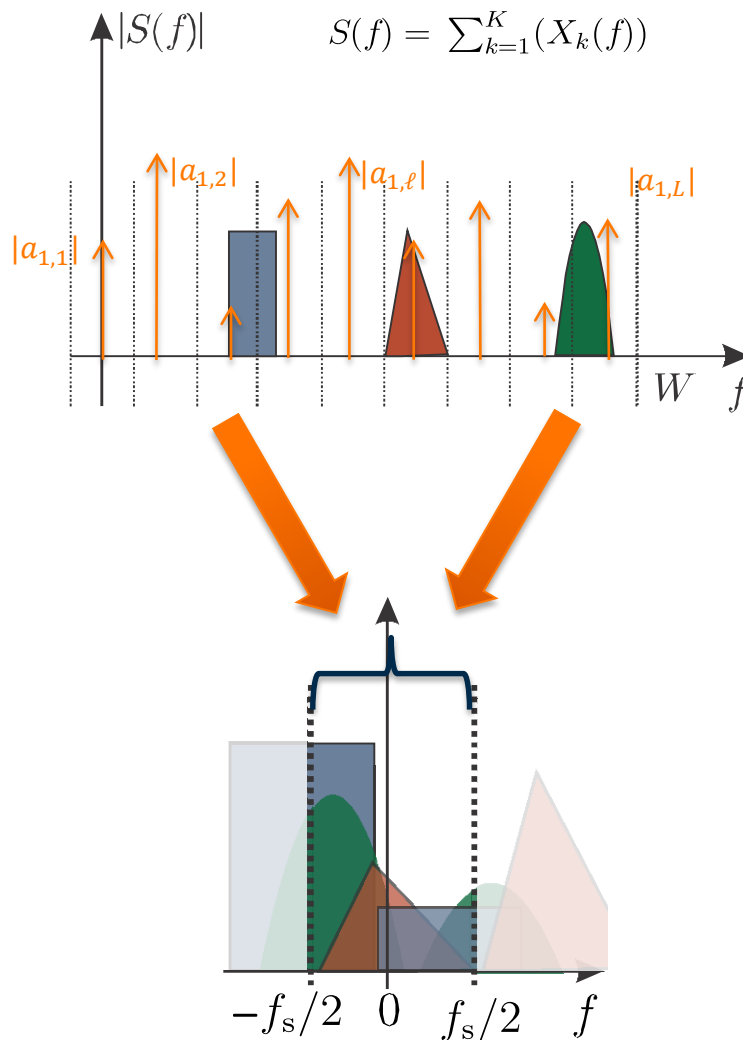
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## Signal sampling

- By means of analog pre-processing alias the individual spectral cells to the baseband with different weights and combine together.

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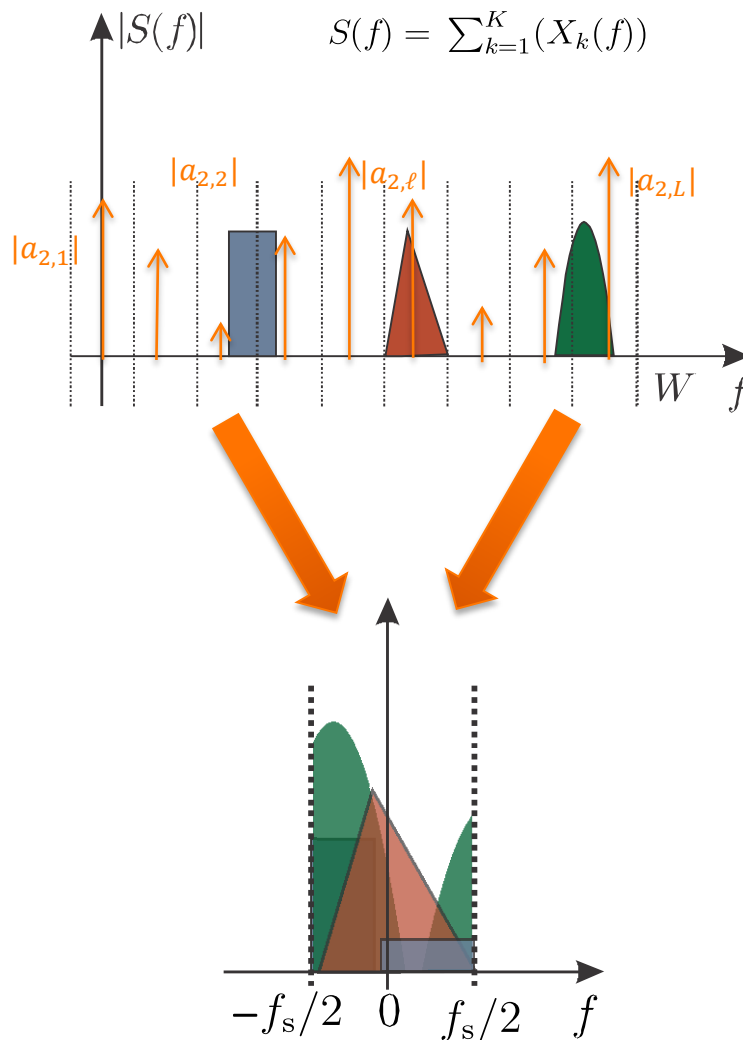
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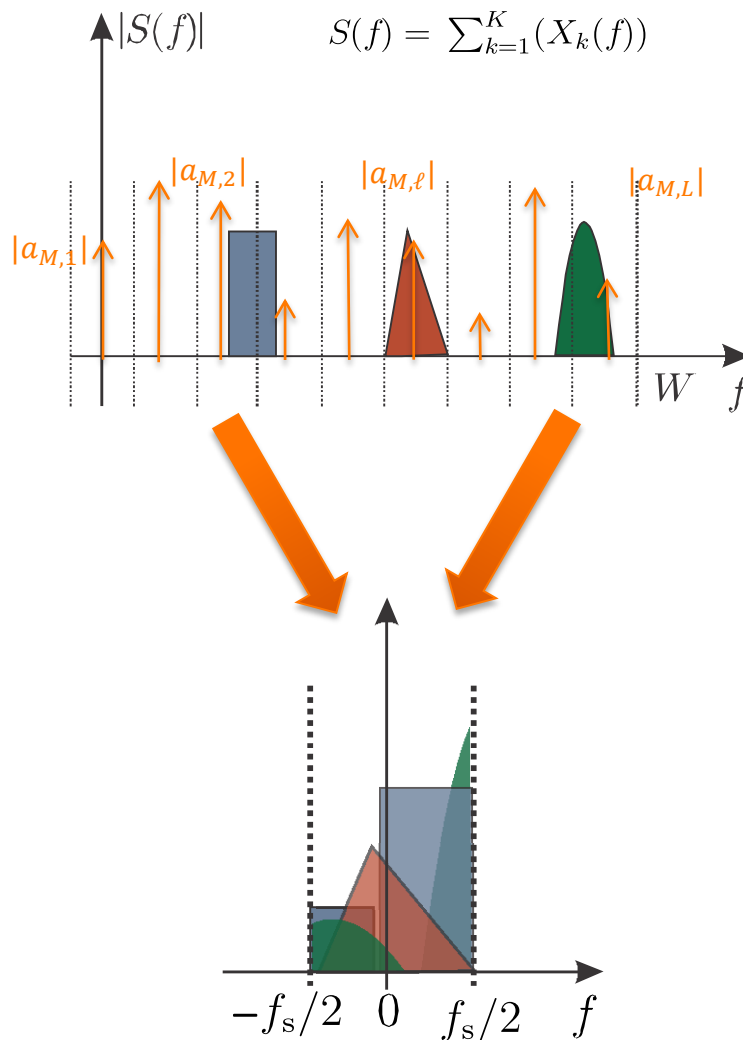
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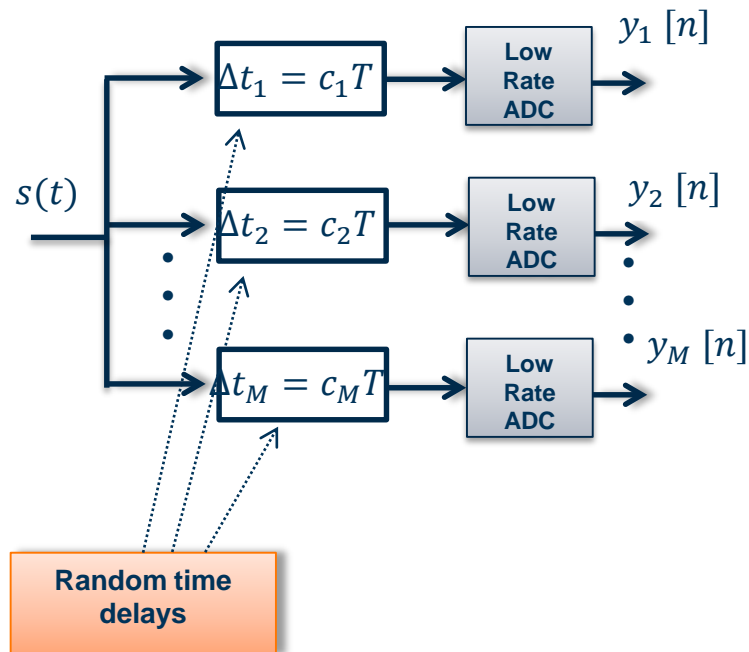
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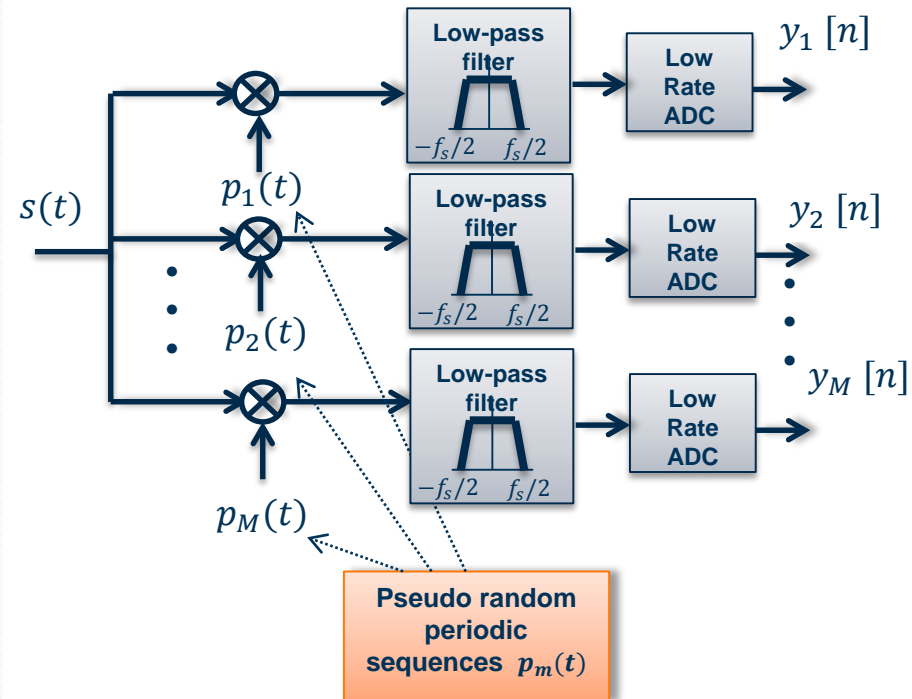
# Sub-Nyquist Signal Acquisition: Sampling Architectures

## Periodic Non-Uniform Sampling<sup>'</sup>



<sup>'</sup> Bresler, Y. (2008, January). Spectrum-blind sampling and compressive sensing for continuous-index signals. In *Information Theory and Applications Workshop, 2008* (pp. 547-554). IEEE.

## Modulated Wideband Converter<sup>\*</sup>

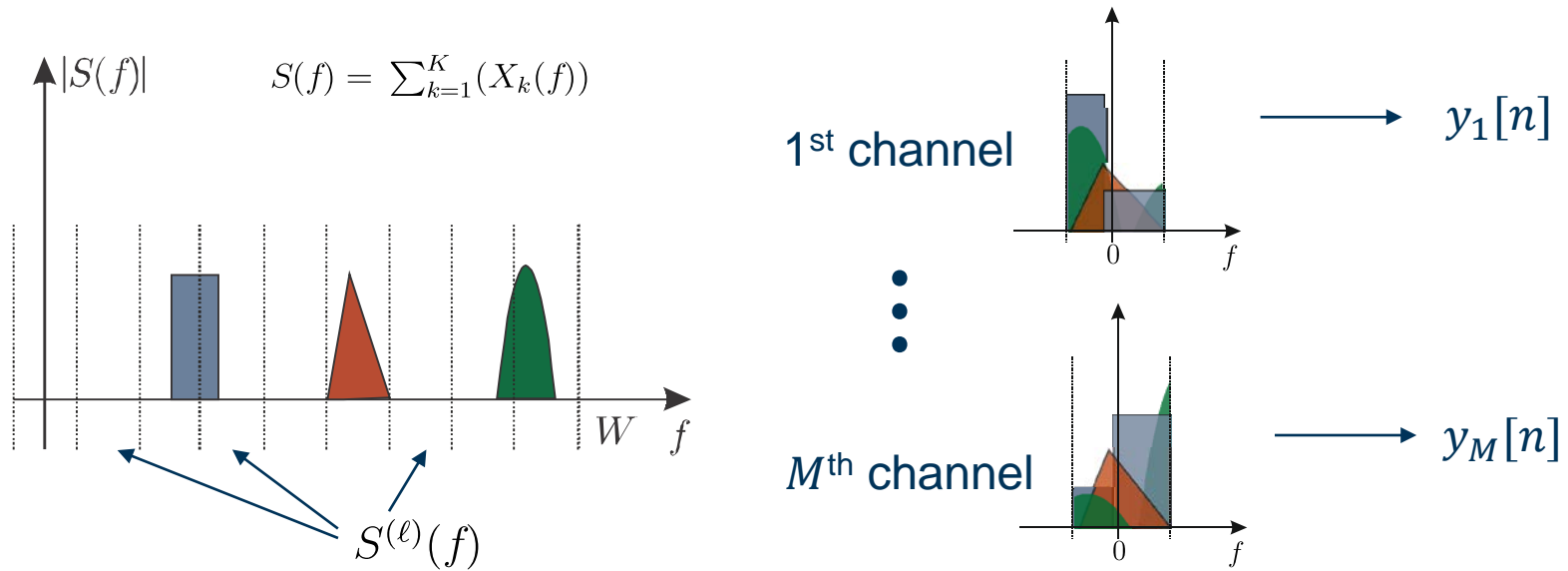


<sup>\*</sup> Mishali, M., & Eldar, Y. C. (2010). From theory to practice: Sub-Nyquist sampling of sparse wideband analog signals. *Selected Topics in Signal Processing, IEEE Journal of*, 4(2), 375-391.

# Outline

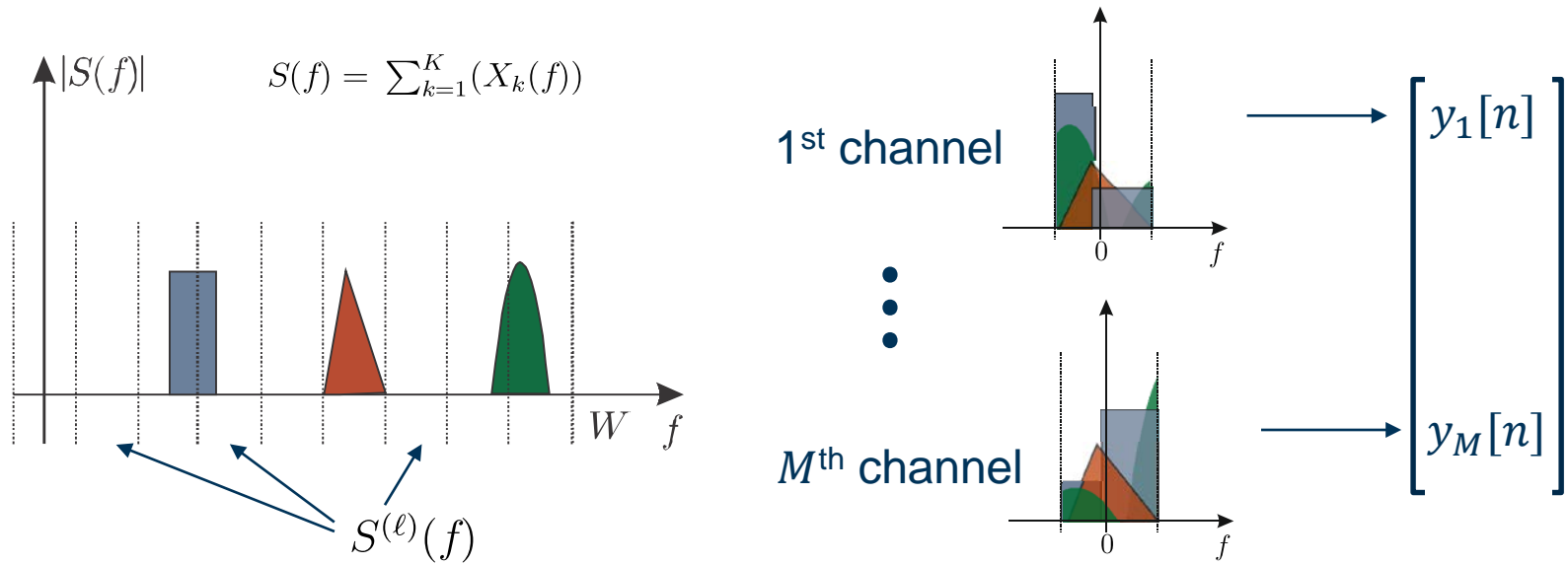
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# Signal recovery





# Signal Recovery



## Output 'compressed' signal

$$\left. \begin{array}{l} \mathbf{y}[n] = \mathbf{A} \cdot \mathbf{z}[n] \\ \text{IDTFT} \Updownarrow \text{DTFT} \\ \mathbf{\dot{y}}(f) = \mathbf{A} \cdot \mathbf{\dot{z}}(f) \end{array} \right\}$$



$$\mathbf{A} = \begin{bmatrix} a_{1,1} & \cdots & a_{1,L} \\ \vdots & \ddots & \vdots \\ a_{M,1} & \cdots & a_{M,L} \end{bmatrix}$$

$$\mathbf{\dot{z}}(f) = \begin{bmatrix} S^{(1)}(f - f_s) \\ \vdots \\ S^{(L)}(f - Lf_s) \end{bmatrix}, f \in \left[-\frac{f_s}{2}, \frac{f_s}{2}\right]$$

# Signal Recovery

- Consider and observation covariance matrix  $\mathbf{R}$

$$\mathbf{R} = \sum_{n=-\infty}^{\infty} \mathbf{y}[n] \cdot \mathbf{y}^H[n] = \int_{f \in \mathcal{F}_s} \dot{\mathbf{y}}(f) \cdot \dot{\mathbf{y}}^H(f), \quad \mathcal{F}_s = [-f_s/2, f_s/2]$$

- Taking into account  $\dot{\mathbf{y}}(f) = \mathbf{A} \cdot \dot{\mathbf{z}}(f)$

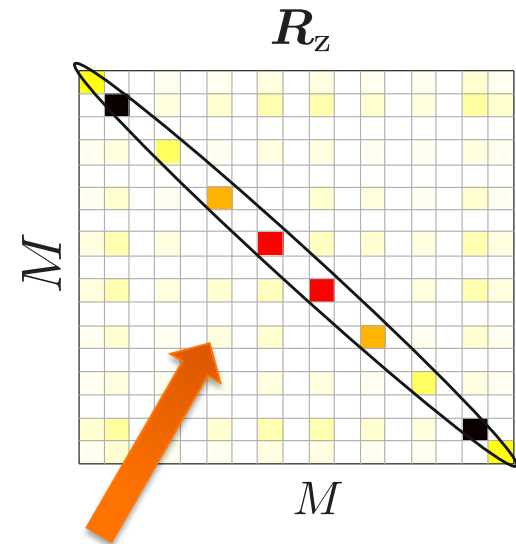
$$\mathbf{R} = \int_{f \in \mathcal{F}_s} \mathbf{A} \cdot \dot{\mathbf{z}}(f) \cdot \dot{\mathbf{z}}^H(f) \cdot \mathbf{A}^H df = \mathbf{A} \cdot \mathbf{R}_z \cdot \mathbf{A}^H$$

where

- diagonal** elements of  $\mathbf{R}_z$  represent the **energy** of the corresponding **spectral cell**

$$(\mathbf{R}_z)_{jj} = \int_{f \in \mathcal{F}_s} |\dot{\mathbf{z}}_j(f)|^2 df = \int_{f \in \mathcal{F}_j} |S^{(j)}(f)|^2 df$$

- non-diagonal** – cross-correlation between spectral cells  $\Rightarrow$  are **equal to 0** assuming that the individual signals are independent from each other

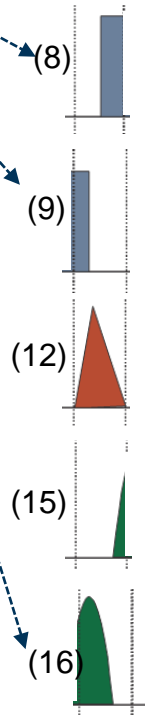
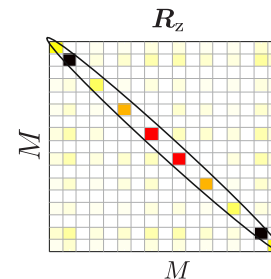


$\tilde{K} \leq 2K$  non-zero elements

# Signal Recovery

- Denote  $\mathcal{S}$  the support of  $\hat{\mathbf{z}}(f)$  – indices corresponding to non-zero  $z_\ell(f) = S^{(\ell)}(f - \ell f)$
- Apply **sparse recovery** approaches from CS to estimated  $\mathcal{S}$  !
- After the positions of the occupied spectral blocks are estimated
  - recover the time samples at the Nyquist rate and perform conventional spectral analysis
  - estimate the PSD directly from the sub-Nyquist measurements
  - directly interpret the results of the support recovery as the coarse spectral occupancy detection

Non-zero elements of the covariance matrix are energies of the occupied spectral blocks



# Outline

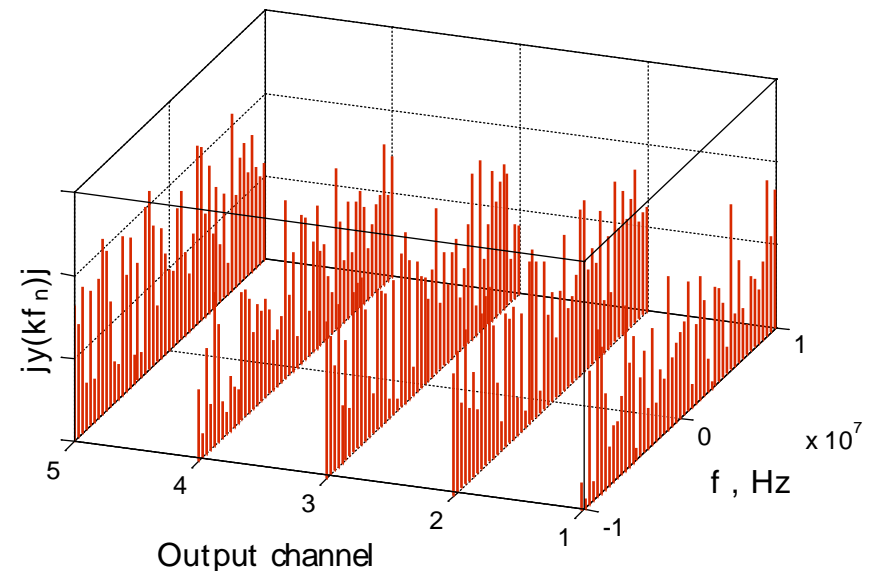
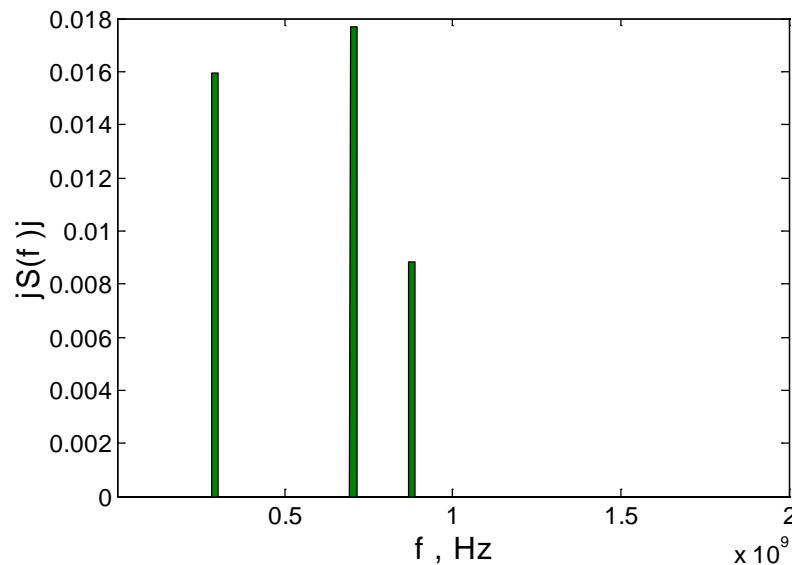
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# Numerical examples. Simulation setup

## Signal and sampling parameters

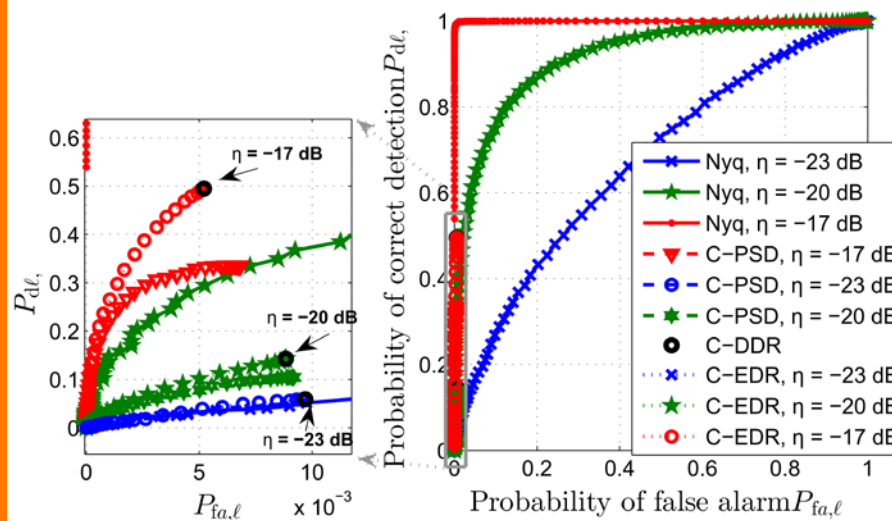
- Test multiband signals were generated in the frequency domain
- The sensing matrix  $\mathbf{A}$  was chosen randomly with entries drawn independently from a complex zero-mean normal distribution

Signal Parameters	Sampling Parameters
$K \in [1, 3]$	$f_{\text{NYQ}}/B = 105$
$B = 19 \text{ MHz}$	$L = 97 \leq f_{\text{NYQ}}/B$
$f_{\text{NYQ}} = 2 \text{ GHz}$	$f_s \approx 20.6 \text{ MHz}$
$N = 50$	$M = 30$

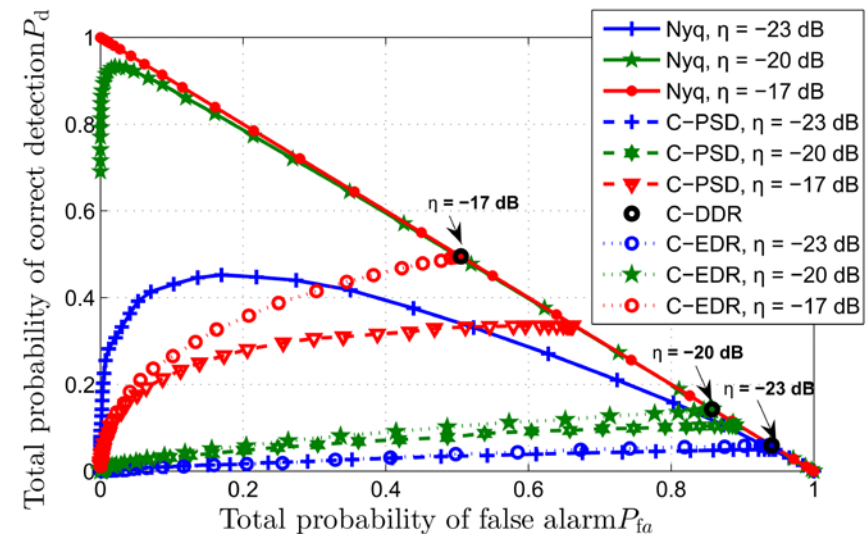


# Numerical results. Single occupied spectral cell

Probability of correct detection per block  
vs. probability of false alarm per block

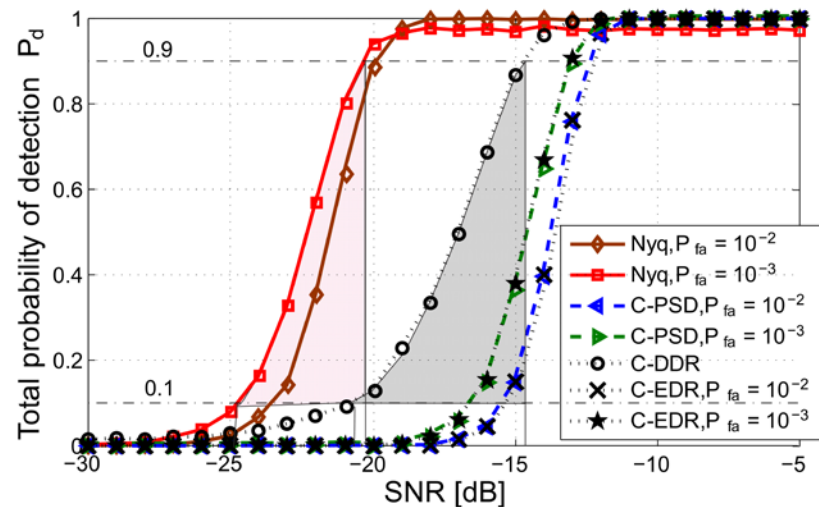


Total probability of correct detection vs.  
total probability false alarm

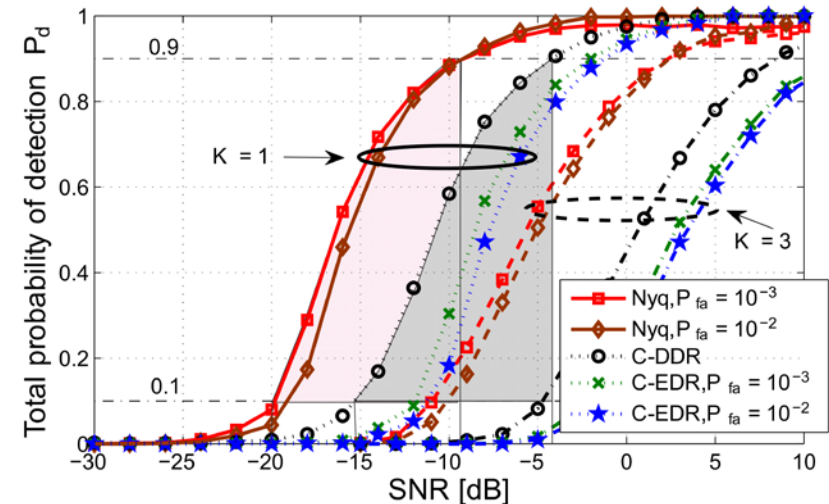


# Numerical results. Single and multiple occupied spectral cells

**Total probability of correct detection vs. SNR for a single occupied spectral cell**



**Total probability of correct detection vs. SNR with no restrictions on the signal sub-band positions**



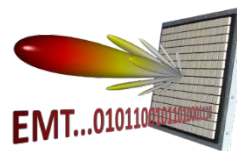
# Conclusions

- We've shown
  - how wideband spectrum sensing can be performed at the **sub-Nyquist** sampling rates.
  - that it is **possible** to infer the **coarse spectral occupancy** directly from the results of the **support** recovery, i.e., without the need to recover full signal, its spectrum or PSD.
- Sub-Nyquist wideband spectrum sensing provides a viable approach to '**fast and dirty**' estimation of the spectral occupancy over wide frequency ranges in highly underutilized environments.



# Thank you for your attention!

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# Spectral Recovery

- Denote  $\mathcal{S}$  the support of  $\mathbf{z}(f)$  – indices corresponding to non-zero  $z_\ell(f) = S^{(\ell)}(f - \ell f)$

- Construct a frame  $\mathbf{V}$  such that

$$\mathbf{V} \cdot \mathbf{V}^H = \hat{\mathbf{R}} = \sum_{n=1}^{N: N > 2K} \mathbf{y}[n] \cdot \mathbf{y}[n]^T$$

- Solving  $\mathbf{V} = \mathbf{A} \cdot \mathbf{U}$  for sparsest  $\mathbf{U}$  yields a matrix with support  $\mathcal{S}$  !

- Once support  $\mathcal{S}$  is found,

$$\begin{cases} z_{\mathcal{S}}[n] = \mathbf{A}_{\mathcal{S}}^\dagger \cdot \mathbf{y}[n] \\ z_i[n] = 0, i \notin \mathcal{S} \end{cases}$$



block sparse recovery

- Generate Nyquist rate sampled sequences and calculate DFT
- Apply DFT to  $\mathbf{z}[n]$  and reshape it into one vector



# PSD Recovery

- Denote  $P_s(f) = \int_{-\infty}^{\infty} \mathbb{E}\{s(t) \cdot s(t-\tau)\} e^{-j2\pi f\tau} d\tau$
- If  $s(t)$  is wide-sense stationary, then  $\mathbf{R}_z(f) = \mathbb{E}\{\mathbf{z}(f) \cdot \mathbf{z}^H(f)\}$  is diagonal and

$$R_{z(i,i)}(f) = P_s(f + (i-1)f_s)$$

- Now  $\mathbf{R}_y(f) = \mathbb{E}\{\mathbf{y}(f) \cdot \mathbf{y}^H(f)\} = \mathbf{A} \cdot \mathbf{R}_z(f) \cdot \mathbf{A}^H$   
and

$$\text{vec}(\mathbf{R}_y(f)) = (\mathbf{A}' \otimes \mathbf{A}) \text{vec}(\mathbf{R}_z(f)) = (\mathbf{A}' \odot \mathbf{A}) \text{diag}(\mathbf{R}_z(f))$$

$\uparrow$  Kronecker                      &                       $\uparrow$  Khatri-Rao  
products

- Construct a frame
- Recover support  $\mathcal{S}$  by solving a finite dimensional sparse recovery problem
- Invert  $\mathbf{A}'_{\mathcal{S}} \odot \mathbf{A}_{\mathcal{S}}$  and reconstruct the power spectral densities  $R_{z(i,i)}$

# Energy Recovery

Recap:

$$\mathbf{R} = \int_{f \in \mathcal{F}_s} \mathbf{A} \cdot \dot{\mathbf{z}}(f) \cdot \dot{\mathbf{z}}^H(f) \cdot \mathbf{A}^H df = \mathbf{A} \cdot \mathbf{R}_z \cdot \mathbf{A}^H$$

- **diagonal** elements of  $\mathbf{R}_z$  represent the **energy** of the corresponding **spectral cell**

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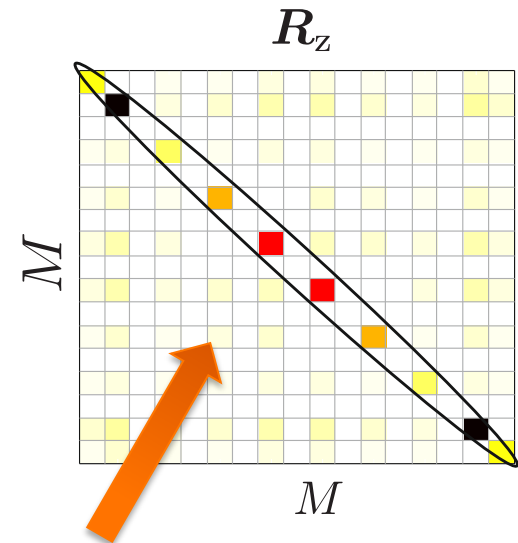
- Let's vectorize the covariance matrix  $\mathbf{R}$

$$\mathbf{r} = \text{vec}(\mathbf{R}) = (\mathbf{A}' \otimes \mathbf{A}) \cdot \mathbf{r}_z = (\mathbf{A}' \odot \mathbf{A}) \cdot \text{diag}(\mathbf{R}_z)$$

$\uparrow$                        $\uparrow$   
 Kronecker      &      Khatri-Rao  
                          products

- Once the unknown support  $\mathcal{S}$  is recovered, the energy  $\varepsilon_{\mathcal{S}}$  of the occupied cells is given by

$$\begin{cases} \varepsilon_{\mathcal{S}} = \Psi_{\mathcal{S}}^\dagger \cdot \mathbf{r} = (\mathbf{A}'_{\mathcal{S}} \odot \mathbf{A}_{\mathcal{S}})^\dagger \cdot \mathbf{r} \\ \varepsilon_i = 0, \quad i \notin \mathcal{S} \end{cases} \quad \longrightarrow$$

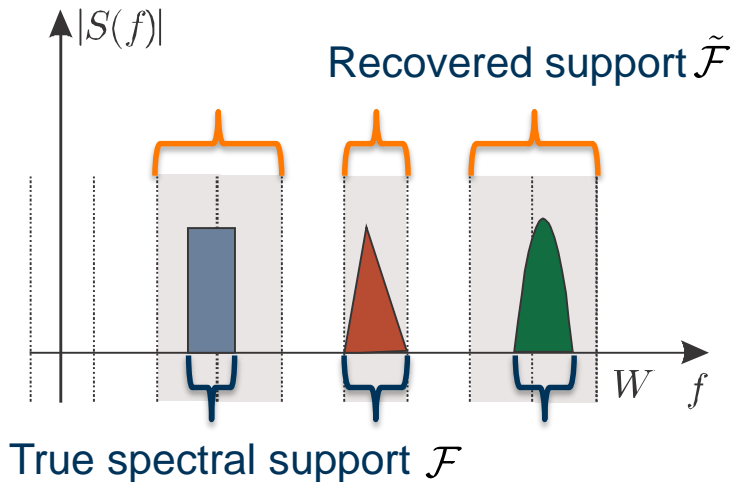


$\tilde{K} \leq 2K$  non-zero elements

The results of the support recovery  
– coarse energy detection at the  
resolution of spectral cells

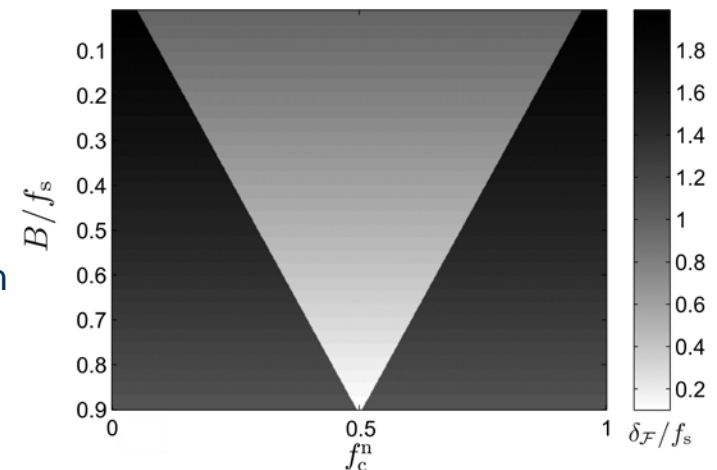


# Wideband Spectrum Sensing



- In the blind case direct compressive ED provides a **coarse estimate** of the spectral occupancy – at the resolution of  $f_s$

## Normalized spectral support difference



- The **difference** between true and recovered spectral support **depends** on the **signal** parameters
- For an **unknown** uniformly distributed random signal spectral support each **signal sub-band** is split between **two** neighbouring **spectral cells** with a high probability

# Numerical examples. Simulation setup

## Performance metrics

- 'Per block' probabilities of false alarm and missed detection

$$P_{\text{fal},\ell} = Pr(b_\ell = 1 | \int_{\mathcal{F}^\ell} |S^\ell(f)|^2 df = 0)$$

$$P_{\text{mdl},\ell} = Pr(b_\ell = 0 | \int_{\mathcal{F}^\ell} |S^\ell(f)|^2 df > 0)$$

- Total probabilities of correct detection, false alarm and missed detection

$$P_d = Pr(b_\ell = 1 \forall \ell \in \mathcal{S}, b_k = 0 \forall k \in \mathcal{S}')$$

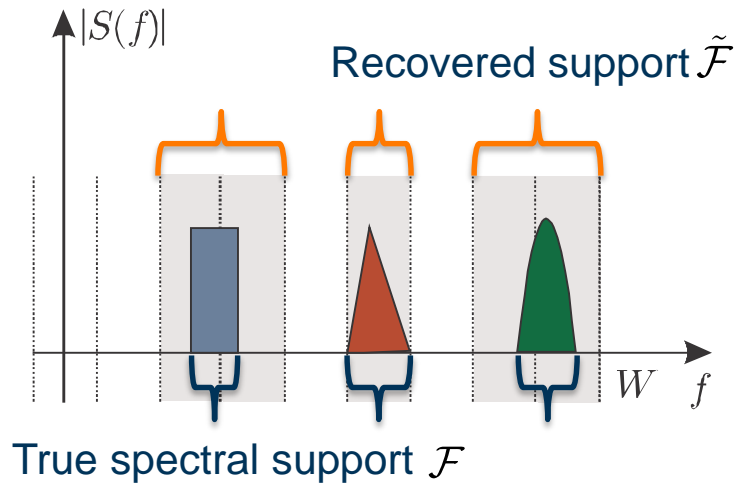
$$P_{\text{fa}} = Pr(\exists k \in \mathcal{S}': b_k = 1)$$

$$P_{\text{md}} = Pr(\exists k \in \mathcal{S}: b_k = 0)$$

## Evaluated multichannel ED approaches

- Power spectral density (PSD) based:
  - PSD recovered from **R – C-PSD**
  - PSD estimated from a Nyquist signal equivalent (no compression) – **Nyq**
- Compressive direct energy detection
  - direct decision rule – **C-DDR**
  - energy based decision rule – **C-EDR**
    1. first estimate the energy within the spectral cell
    2. apply classical ED

# Wideband Spectrum Sensing



- In the blind case direct compressive ED provides a **coarse estimate** of the spectral occupancy – at the resolution of  $f_s$

$$0 \leq \delta_f = \lambda(\tilde{\mathcal{F}}) - \lambda(\mathcal{F}) = \tilde{K} f_s - \sum_i B_i \leq 2K f_s$$